

NUMERICAL SIMULATIONS OF INNOVATIVE GROUND PLANE AND DOUBLE-GATE CONFIGURATIONS IN THIN-BODY AND – BURIED OXIDE OF SOI MOSFETs

by

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A thesis submitted in fulfillment of the requirements for the degree of Doctor of Philosophy

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LIST OF ABBREVIATIONS

1DEG One-dimensional electron gas

2D Two-dimensional

2DEG Two-dimensional electron gas

3D Three-dimensional ACAlternating current

ADG Asymmetric double-gate

Buried oxide BOX

CCConstant current

nal copyright Complementary metal-oxide-semiconductor **CMOS**

DC Direct current

DG Double-gate

DIBL Drain-induced barrier lowering

ΕI Electrostatic integrity

ELR Extrapolation of linear region

ENG Effective number of gates

EOT Equivalent oxide thickness

Extremely thin SOI **ETSOI**

Floating body effect **FBE**

Fully depleted FD

Fully-depleted silicon-on-insulator FD-SOI

Fin field-effect transistor **FinFET**

FoM Figure-of-merit GAA Gate-all-around

GMLE Transconductance linear extrapolations

GP Ground plane

High permittivity gate-dielectric High- κ

HRTEM High-resolution transmission electron microscopy

IC Integrated circuit

International Technology Roadmap for Semiconductor **ITRS**

Low permittivity gate-dielectric Low-κ

MOSFET Metal-oxide-semiconductor field-effect transistor NMOS N-type metal-oxide-semiconductor

PD-SOI Partially-depleted silicon-on-insulator

PMOS P-type metal-oxide-semiconductor

QDG Quasi double-gate
RF Radio frequency
RM Ratio method

SCE Short-channel effects

S/D Source/drain

SD Second derivative

SDL Second derivative logarithmic

SG Single-gate

SiGe Silicon germanium
Si-SiO₂ Silicon-silicon dioxide

SiO₂ Silicon dioxide SiN Silicon nitride

SOD Silicon-on-diamond

SOI Silicon-on-insulator
SON Silicon-on-nothing
SRH Shockley-Read Hall
SS Subthreshold swing

TCAD Technology Computer Aided Design

TEM Transmission electron microscopy

UTB Ultra-thin body

UTBB Ultra-thin body and BOX

LIST OF SYMBOLS

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A Area

 $A_{\rm v}$ Intrinsic gain C Capacitance

 $C_{\rm gg}$ Gate-to-gate capacitance

 $C_{\text{ox}1}$ Front gate oxide capacitance

 C_{ox2} Buried oxide capacitance C_{si} Silicon film capacitance

*E*_i Intrinsic Fermi level

f Frequency

 $f_{\rm t}$ Current gain cut-off frequency

 $g_{\rm d}$ Output conductance $g_{\rm m}$ Transconductance

 $g_{\rm m \ max}$ Maximum transconductance

 $g_{\text{m} \text{ max norm}}$ Normalized maximum transconductance

I Current

*I*_d Drain current

 $I_{\rm d}$ - $V_{\rm g}$ Plot of drain current versus gate voltage

 $I_{\rm d_norm}$ Normalized drain current

 $I_{
m on}$ On-state current $I_{
m off}$ Off-state current I-V Current voltage

k Boltzmann constant

 κ Gate dielectric permittivity

 $\kappa_{\text{high-}\kappa}$ Permittivity of high- κ gate dielectric

 κ_{SiO2} Permittivity of silicon dioxide gate dielectric

 $L_{\rm el}$ Electrical gate length

 $L_{\rm g}$ Gate length n Body factor

 n_i Intrinsic carrier concentration N_A Acceptor doping concentration

 $N_{\rm D}$ Donor doping concentration

P Power

q Charge of an electron

T Temperature

 $t_{\text{high-}\kappa}$ Thickness of high- κ gate dielectric

 $T_{\rm BOX}$ Buried-oxide (BOX) thickness

 T_{dep} Depletion layer thickness

 $T_{\rm ox}$ Gate oxide thickness

 $T_{\rm ox\ el}$ Electrical oxide thickness

 $T_{\rm si}$ Silicon-body thickness

 $V_{\rm bi}$ Built-in voltage $V_{\rm d}$ Drain voltage

 $V_{\text{drain (llinear)}}$ Linear drain voltage

 $V_{\text{drain(saturation)}}$ Saturation drain voltage

 $V_{\rm dd}$ Supply voltage $V_{\rm g}$ Gate voltage

 $V_{\rm th}$ Threshold voltage

 $V_{\text{th(linear)}}$ Linear threshold voltage

 $V_{\text{th(saturation)}}$ Saturation threshold voltage

 V_{sub} Substrate voltage W Width of channel

€ Electric field

 ε_{ox} Permittivity of oxide

 ε_{Si} Permittivity of silicon

 ε_{SiO2} Permittivity of silicon dioxide

 x_{dmax} Depletion zone extending from the Si-SiO₂ interface to the maximum

depletion width

 x_{j} Junction depth

au Gate delay

μ Mobility

 λ Fitting parameter to take into account the contribution of BOX

fringing field

 α Scaling factor

 $\Phi_{\rm F}$ Fermi level

 $\Phi_{\rm m}$ Gate workfunction $\psi_{\rm Horizontal}$ Horizontal potential $\psi_{\rm s}$ Surface potential $\psi_{\rm Vertical}$ Vertical potential

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Simulasi Berangka bagi Konfigurasi Inovatif Satah-Bumi dan Get-Berkembar dalam SOI MOSFETs Badan dan -Oksida Tertanam Nipis

ABSTRAK

Penskalaan transistor membolehkan peningkatan dalam ketumpatan transistor, kelajuan pensuisan dan kekompleksan dengan tiada peningkatan dalam penggunaan kuasa. Walaubagaimanapun, penskalaan transistor MOS yang lazim nampak menuju ke arah akhir pelan tindakan teknologi disebabkan oleh kebolehubahan prestasi yang semakin buruk dan kesan saluran-pendek (SCEs). Salah satu pesaing yang dijangka menggantikan seni bina transistor semasa adalah SOI MOSFETs planar badan dan oksida tertanam nipis (UTBB). Kelebihan struktur SOI badan-nipis terletak pada proses planarnya yang mudah yang serasi sepenuhnya dengan aliran CMOS silikon pukal. Dalam kerja kajian ini, perhatian khusus diberikan terhadap prestasi UTBB SOI MOSFETs dengan BOX yang nipis dalam meningkatkan tingkah laku elektrostatik oleh badan-nipis dibandingkan dengan transistor SOI dengan BOX yang tebal (UTB) untuk melanjutkan kebolehskalaan CMOS. Selanjutnya, UTBB dengan seni bina satah bumi (GP) dan konfigurasi get yang berbeza (i.e. get-tunggal (SG) dan get-ganda (DG)) dikaji secara menyeluruh melalui simulasi berangka sebagai calon yang mungkin untuk meneruskan Hukum Moore. Kajian mendalam mengenai angka merit (FoM) digital dan analog/RF dijalankan dalam julat frekuensi yang lebar (dari 0.01 Hz sehingga 100 GHz) dalam hubungan dengan mekanisme operasi peranti. Didapati bahawa pembentukan GP inovatif yang terdiri daripada GP setempat jenis -p dalam substrat di bawah saluran (di sini dirujuk sebagai GP-B dalam tesis) menekan kesan susutan substrat secara efektif dan menunjukkan imuniti yang lebih baik terhadap SCEs daripada pandangan analisis digital. Peningkatan selanjutnya dalam imuniti terhadap SCEs dapat dicapai dengan konfigurasi DG di mana kesan seni bina GP yang berbeza digandakan dibandingkan dengan SG. Walaupun penggunaan konfigurasi DG memberikan prestasi digital yang unggul, nilai frekuensi potongan gandaan intrinsik (f_t) adalah rendah dalam domain analog berbanding SG disebabkan oleh peningkatan kemuatan berparasit get-ke-get (C_{gg}) . Oleh itu, pemilihan yang cermat dan keseimbangan diperlukan apabila memilih struktur peranti tertentu di mana hasil yang diperolehi daripada penyelidikan ini menyumbang kepada pengenalpastian seni bina GP dan konfigurasi get (SG atau DG) yang boleh dipakai dalam rekabentuk peranti untuk disesuaikan dengan aplikasi spesifik sama ada digital atau RF.

Numerical Simulations of Innovative Ground Plane (GP) and Double-gate (DG) Configurations in Thin-body and –buried Oxide of SOI MOSFETs

ABSTRACT

The downscaling of transistors enables an increased in transistor density, faster switching speeds and greater complexity with no increase in power consumption. However, the scaling of the conventional planar MOS transistors appears to be reaching the end of the technology roadmap due to worsening performance variability and shortchannel effects (SCEs). One of the contenders anticipated to replace the current transistor architecture is planar ultra-thin body and BOX (UTBB) SOI MOSFET. The advantage of the thin-body SOI structure lies in its simple planar process which is fully compatible with the bulk silicon CMOS flow. In this research work, a particular attention is being given to the performance of UTBB SOI MOSFETs with its thin BOX in improving electrostatics behaviour namely of drain-induced barrier lowering (DIBL) of the thin-body as compared to thick BOX (UTB) SOI transistors for extending CMOS scalability. Subsequently, UTBB with different ground plane (GP) architectures and gate configurations (i.e. single-gate (SG) vs double-gate (DG)) are extensively studied through numerical simulations as possible candidates for the continuation of Moore's Law. In-depth study of the digital and analog/RF figure-of-merit (FoM) are carried out in a wide range of frequency (from 0.01 Hz to 100 GHz) in correlation with device operation mechanisms. It is discovered that an innovative GP formation made of localized GP of p-type in the substrate underneath the channel (referred herein throughout the thesis as GP-B) effectively suppress substrate depletion effects and shows better immunity against SCEs from the digital analysis viewpoint. Further improvements in the immunity against SCEs can be achieved in DG configurations where the impact of different GP architectures is amplified as compared to SG. Even though the use of DG configurations provides superior digital performance, lower current gain cut-off frequency (f_t) values are produced than SG in the analog domain due to an increase of gate-to-gate capacitances (C_{gg}). Therefore, careful selections and trade-offs are needed when selecting a particular device structure where the results obtained in this research work contribute to the identifications of GP architectures and gate configurations (SG or DG) that can be adopted in device design to suit specific applications of either digital or RF.

CHAPTER 1

INTRODUCTION

1.1 General review of CMOS Technology

1.1.1 Scaling of CMOS Technology to Their Limits

The integrated circuit (IC) technology has followed Moore's Law since 1965 which states that the number of devices integrated double every 18 months. This progression is made possible by continuous miniaturization in feature size of components devices which are integrated a concept known as device scaling. In detailed, device scaling refers to scaling of various structural parameters of a MOSFET to ensure the device continue to function properly. These include lateral as well as vertical dimensions such as the channel length, the width, the source/drain junction depth (x_i) and the gate oxide thickness (T_{ox}) . For proper device scaling, power supply voltages should also be reduced to keep the internal field constant. The first complete scaling scheme known as constant-field scaling was introduced by Dennard et al. (1974) as shown in Table 1.1 and is regarded as the seminal reference in scaling theory for MOSFET integrated circuits. Depending on the variable, the parameter could be multiplied, or divided by α which is a unitless scaling factor. However, as voltage is not usually scaled as fast as the linear dimensions due to subthreshold leakage constraint, additional scaling factor for the electric field (ϵ) is introduced to account for the increased of ϵ and is summarized as 'generalized scaling factor' as shown in Table 1.1.

The scaling of the transistor's feature size leads to an increased speed and improved density (smaller areas for devices and circuits).

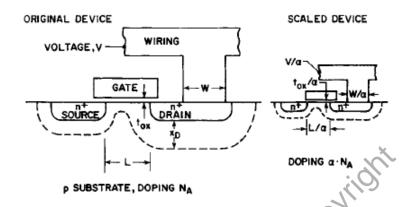


Figure 1.1: Principles of MOSFET constant-field scaling (Davari, Dennard, & Shahidi, 1995)

Table 1.1: The classical sealing trends

Parameters	Constant-field scaling (Dennard et al., 1974)	Generalized scaling factor (Baccarani, Wordeman, & Dennard, 1984)
Physical dimensions (L_g, W, T_{ox}, x_j)	1/α	$1/\alpha$
Electric field (€)	1	ϵ
Body doping concentration (N_A)	α	ϵ / α
Supply voltage (V_{dd})	1/lpha	$\epsilon \alpha$
Transistor current (I)	$1/\alpha$	ϵ/α
Capacitance	$1/\alpha$	1/lpha
$(C = \varepsilon_{ox}A/T_{ox})$		
Area (A)	$1/\alpha^2$	$1/\alpha^2$
Gate delay	$1/\alpha$	ϵ/α
($\tau \sim C V_{\rm dd} / I$)		
Power dissipation	$1/\alpha^2$	ϵ^2/α^2
$(P \sim IV_{\rm dd})$		
Power density (P/A)	1	ϵ^2

1.1.2 Scaling Challenges

The classical scaling technique of MOSFET was followed successfully until 90 nm transistor generation (M. Bohr, 2008, 2009; Kuhn, 2009) (The last CMOS generation where the downscaling of transistor to make it smaller is adequate to improve the transistor's performance is of the 130 nm transistor generation). In the following generation, it is then recognized that simple scaling of bulk MOSFETs i.e. increasing the doping in the channel and reducing the silicon thickness is no longer valid as a result of rapidly increasing random variability and poor short channel immunity. With the shrinking of the transistor gate length (L_g), the lateral electric fields at the source and drain can penetrate into the channel, causing reduction in barrier height of source/body junction. This will lead to an increase in short-channel effects (SCEs).

SCEs arise when the close proximity between the source and the drain causes the gate to lose control of the potential distribution and the flow of current in the channel region. With shorter $L_{\rm g}$, the depletion regions of high electric fields associated with the source and drain regions started to interact with each other, causing direct carrier transport between the source and drain. This reduces control of the gate over the channel and in turn, rise in off-state current ($I_{\rm off}$) and lower threshold voltage ($V_{\rm th}$) are observed. In conventional MOSFET, decreasing $L_{\rm g}$ has been accompanied by the decreased in gate oxide thickness ($T_{\rm ox}$) and the source/drain junction depth ($x_{\rm j}$). However, an increased in gate leakage current caused by tunnelling through the very thin oxide (\sim 2 nm) has put the limit to oxide scaling. It is then proposed that higher